(54) Title: ENHANCED ERROR CONCEALMENT FOR SPATIAL AUDIO

Error concealment
Frame error in one channel

(57) Abstract: An error concealment
method for multi-channel digital audio
involves receiving an audio signal
having audio data forming a first audio
channel and a second audio channel
included therein, wherein the first and
second audio channels are correlated
with each other in a manner so that a
spatial sensation is typically perceived
when listened to by a user. Erroneous
first-channel data is detected in the
first audio channel, and second-channel
data is obtained from the second audio
channel. The erroneous first-channel
data of the first audio channel is
corrected by using the second-channel
data. Upon detection of the erroneous
first-channel data, a spatially perceivable
inter-channel relation between the first
and second audio channels is determined,
and the determined inter-channel relation
is used when correcting the erroneous
first-channel data of the first audio
channel so as to preserve the spatial
sensation perceived by the user.

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ENHANCED ERROR CONCEALMENT FOR SPATIAL AUDIO

Field of the Invention

The present invention relates to an error conceal-
ment method for multi-channel digital audio, where an
5   audio signal is received which has audio data forming a
first audio channel and a second audio channel included
therein, said first and second audio channels being
correlated with each other in a manner so that a spatial
sensation is typically perceived when the signal is
10 listened to by a user.

More specifically, the present invention relates to
an error concealment method, where erroneous first-
channel data is detected in the first audio channel,
second-channel data is obtained from the second audio
15 channel, and the erroneous first-channel data of the
first audio channel is corrected by using the second-
channel data.

Background of the Invention

20 Multi-channel audio is used in various applica-
tions, such as high-quality (stereo) music or audio con-
ferencing (teleconferencing), for creating an impression
of the direction of the sound source in relation to the
listener (user). Thus, the multi-channel audio generates
25 a spatial effect. In the case of audio conferencing, the
spatial effect is created artificially by spatialization
in the teleconference bridge, and in the case of stereo
music the effect is created already by the recording (or
mixing) arrangement. Reference will now be made, in an
30 exemplifying manner, to a teleconferencing system in-
cluding not only a teleconference bridge but also, of
course, a plurality of user terminals which are all coup-
led to the teleconference bridge through a communications

network, such as a packet-switched mobile telecommunica-
tions or data communications network. One example of a
teleconferencing system is shown in FIG 3.

The conference bridge 300 is responsible for re-
5   ceiving mono audio streams from microphones 312, 322,
332, 342 of a plurality of user terminals 310, 320, 330
340 and processing these mono streams (in terms of e.g.
automatic gain control, active stream detection, mixing
and spatialization, as well as artificial reverberation)
10   so as to provide a stereo output signal to the user ter-
minals. The user terminals are responsible of the actual
audio capture (through microphones 312, 322, 332, 342)
and audio reproduction (through speaker pairs 314/316,
324/326, 334/336, 344/346). The stereophonic connection
15   from the teleconference bridge to the user terminal makes
it possible to transmit spatial audio which is processed
for headphones or loudspeakers. Sound sources can be
spatialized around the user by exploiting known 3D audio
techniques, such as HRTF filtering ("Head Related Trans-
20   fer Function"). Use of spatial audio improves speech
intelligibility and facilitates speaker detection and
separation. Moreover, it will let the conference en-
vironment sound more natural and satisfactory. The
stereophonic sound can be transmitted as two separately
25   coded mono channels or as one stereo-coded channel.
Alternatively, the stereophonic sound can be transmitted
on one mono channel, e.g., in which channels are
interleaved.

Transmission errors can have harmful side effects
30   in spatial audio conferencing systems. When speech frames
from one or both channels are lost, the perceived spatial
image will very likely shift its location. The shifting
can be very disturbing for the listener. For example, a
sound source, which was spatialized at the listeners left
35   side, may shift rapidly to the center or even to the
other side and back again. Whereas in monophonic con-
ferencing the user would not even notice some frame

errors, the spatial dimension of stereo conferencing will
make these errors all the more noticeable. This is a
problem especially in cases where two independent in-
stances of a speech codec originally designed for single
5   channel are used to process two channels of a stereo
conference.

If frames are corrupted or even lost during trans-
mission, the speech decoder typically uses error con-
cealment (masking) methods that are based on extra-
10 polation to substitute the erroneous or missing frames in
the output speech signal. The extrapolation is based on
previous frames in the same channel. This sort of error
concealment is designed for monophonic signals and does
not generally perform well with spatialized signals. When
15 such known single-channel based error concealment methods
are used for spatialized multi-channel audio, the result
can be a shifting spatial image during frame errors.
Stationary spatial image requires that the phase diffe-
rences between the signal components of the channels be
20 preserved in all circumstances. Single-channel based
error concealment methods cannot guarantee that the
extrapolated signals are correctly phase-shifted (and
linearly filtered) copies of each other.

Typically, the worst case would occur during un-
25 voiced phonemes, where extrapolation can not preserve the
phase difference between the channels, because generally
the error concealment methods cannot reconstruct the
details of the missing unvoiced (noise-like) signal as
accurately as would be required to preserve the desired
30 phase difference. The reason for this is that an unvoiced
signal typically resembles white noise and cannot there-
fore be effectively predicted/extrapolated. In a single-
channel case, good error concealment performance for an
unvoiced signal can be reached simply by generating a
35 noise-like signal with constant or smoothly changing
energy level. Unfortunately, however, this does not work
well for a spatialized multi-channel signal, since the

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correlation between the channels (that are processed
independently of each other) would be lost.

The nature of the errors depends on the trans-
mission system. The errors occurring in an over-the-air
5   connection typically differ from buffer overflow to er-
rors in network routers. The errors may appear as single
frame errors or error bursts where typically several
consecutive frames are lost. The transmission system also
determines whether the channels are transmitted and
10   possibly routed independently from each other or as a
common "interleaved" channel. Thus, speech frames may be
lost at one of the channels only but also simultaneously
at both of the channels.

If the error affects only one of the channels, an
15   attempt can be made to correct the error by using data
from the unaffected other channel. In such a case, if the
other channel is simply copied over the erroneous chan-
nel, this replacement operation would set the actual
phase difference between the signals to zero for the er-
20   roneous frame. For example, if the sound source is spati-
alized to the listener's side when there is an interaural
time difference, as a result of the replacement opera-
tion, the directional impression of the sound would be
quickly lost by the human auditory system. Already a few
25   successive monophonic samples are perceived as a change
of the sound position. In addition, for a voiced
(periodic) signal, the replacement operation would also
introduce discontinuity in the periodic signal structure.
Because human hearing is very sensitive to such
30   discontinuities, annoying clicks will often be heard at
boundaries of actually received and replaced parts of the
signal, and the sound source will appear to move towards
the center. Therefore, this method does not work for
spatial audio.

35   To avoid shift in spatial position during frame
loss, it is previously known to fade both channels out
(muting). This, however, has a drawback in that a silent

gap will be caused whenever an error is detected. A
silent gap during continuous speech will be disturbing,
especially when there is noticeable background noise at
the remote speaker location.

5   US-6 351 727 and US-6 351 728 present various error
concealment methods in line with those described above,
such as muting, left-right substitution, repeating and
estimating. Instead of merely performing the error con-
cealment upon an entire audio frame in which an error has
10   been detected during decoding, US-6 351 727 and US-6 351
728 suggest replacing only those portions which actually
contain errors. These portions of an audio frame may be
certain (groups of) spectral values or sub-bands, in-
cluding time domain sample values or spectral domain
15   sample values. The error concealment may also be based on
certain parameters from the decoder, such as scale fac-
tors or other control data,

Summary of the Invention

20   In view of the above, an objective of the invention
is to solve or at least reduce the problems discussed
above. In more particular, a purpose of the invention is
to prevent a shift in the spatial position of a sound
source when concealing frame errors. Thus, the invention
25 seeks to preserve the spatial position or sensation which
is perceived by a user when listening to multi-channel
audio, even if errors are generated and corrected by
inter-channel use of audio data.

      Generally, the above objectives are achieved by an
30   error concealment method, a receiver of multi-channel
digital audio, a computer program product, an integrated
circuit, a user terminal and a teleconference system
according to the attached independent patent claims.

      Simply put, to achieve the above, the invention
35   exploits co-channel redundancy of spatial audio for error
concealment. If an audio frame is corrupted or missing in
one of the channels, audio data from the correctly re-

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ceived channel is used to reconstruct the erroneous
frame. The invention may be exercised in time domain,
wherein an inter-channel time difference or phase dif-
ference between the channels is determined and used when
5 reconstructing the erroneous frame, so that the perceived
spatial position of the sound source is preserved. The
invention may also be exercised in "parameter domain", in
conjunction with a speech codec known per se, wherein an
erroneous frame on one channel is reconstructed from
10 parameter data of previous frames on that channel but
also from parameter data of a concurrent non-erroneous
frame on the other channel.

If audio frames are simultaneously corrupted or
missing in both of the channels, to prevent a shift of
15 sound source location, the audio signal for one channel
is first reconstructed by using an error concealment
method known per se (such as extrapolation from preceding
frames in that channel), and then the other channel is
reconstructed based on the first reconstructed channel in
20 the manner described above.

One advantage of the invention is that not only a
single erroneous frame but also longer sequences of
consecutive erroneous frames may be corrected without
significant loss of perceived spatial position.
25   A first aspect of the invention is an error con-
cealment method for multi-channel digital audio. The
method comprises the steps of

receiving an audio signal having audio data forming
a first audio channel and a second audio channel included
30 therein, said first and second audio channels being cor-
related with each other in a manner so that a spatial
sensation is typically perceived when listened to by a
user;

detecting erroneous first-channel data in the first
35 audio channel;

obtaining second-channel data from the second audio
channel; and

correcting the erroneous first-channel data of the
first audio channel by using the second-channel data;
        as well as the steps of
        determining, upon detection of the erroneous first-
5   channel data, a spatially perceivable inter-channel
relation between the first and second audio channels; and
        using the determined inter-channel relation when
correcting the erroneous first-channel data of the first
audio channel so as to preserve the spatial sensation
10   perceived by the user.
        The erroneous first-channel data of the first audio
channel may be corrected by manipulating the second-
channel data in accordance with the determined inter-
channel relation and then replacing the erroneous first-
15   channel data with the manipulated second-channel data.
        In one embodiment, the determined inter-channel
relation may be a phase difference between the first and
second audio channels. The manipulation of the second-
channel data may then consist in selecting the second-
20   channel data from the second audio channel with a time
shift with respect to the first audio channel, said time
shift corresponding to the determined phase difference.
        The phase difference may be determined by analyzing
the first and second channels of the received audio sig-
25   nal with respect to each other. This analysis may involve
calculating the cross-correlation between the channels.
It may alternatively involve low-pass filtering of each
of the first and second channels, and detecting the
phases of the first and second channels after low-pass
30   filtering by matching peaks and/or zero-crossings in
voiced phonemes.
        In another embodiment, the determined inter-channel
relation or phase difference may be determined from meta-
data received together with the audio signal.
35   The method may involve an additional step of de-
coding the received audio signal prior to detecting er-
roneous first-channel data in the first audio channel.

The first and second audio channels may each com-
prise a plurality of audio frames, and the detection and
correction of erroneous first-channel data may concern at
least one entire audio frame. Alternatively, the detec-
5   tion and correction of erroneous first-channel data may
concern only part(s) of an audio frame, such as certain
spectral sub-band(s), or even parts thereof.

As yet another alternative, the detection and cor-
rection of erroneous first-channel data may concern only
10   some audio component(s), such as principal audio compo-
nent(s), which is/are detected or indicated to be present
in the audio signal.

The detection and correction of erroneous first-
channel data may be performed in the time domain upon a
15   plurality of time domain audio samples contained in the
audio frame.

The first and second audio channels may be left and
right stereo channels, or vice versa. The first and se-
cond audio channels may also be any correlated channels
20   of a 4.1, 5.1 or 6.1 digital audio format, or any other
so-called 3D or spatial audio format, or in general any
two channels which carry audio information and are tem-
porally highly correlated, i.e., derived essentially from
the same sound source.

25   In one embodiment, capable of concurrent error con-
cealment for both channels, the method may comprise the
additional steps, after detecting erroneous first-channel
data in the first audio channel, of:

detecting erroneous second-channel data in the se-
30   cond audio channel, essentially concurrent with the er-
roneous first-channel data detected in the first audio
channel;

selecting either the first audio channel or the
second audio channel as source channel for audio re-
35   construction;

reconstructing the erroneous data of the selected source channel from preceding data in the selected source channel; and

reconstructing the erroneous data of the other of
5   the first and second audio channels, which was not selected as source channel, from the reconstructed data of the source channel in the manner described above.

In this embodiment, the one of the first audio channel or the second audio channel which has the highest
10 signal energy or power level, or alternatively the one which is leading in terms of phase, may be selected as source channel. In the reconstruction, it may then be necessary either to buffer the data from the source channel to obtain a full frame to the other channel, or
15 to buffer the data obtained to the other channel before encoding.

The step of reconstructing the erroneous data of the selected source channel from preceding data in the selected source channel may be performed by attenuated
20 extrapolation or copying of the preceding data.

After having reconstructed the first and second audio channels, the reconstructed audio data may be attenuated, and the first and second audio channels may be maintained attenuated for as long as there are con-
25 secutive errors on the first and second audio channels Then, upon detecting that there are no more consecutive errors on the first and second audio channels, the first and second audio channels may be amplified to cancel the attenuation thereof.

30   The audio signal may be received from a teleconference bridge over at least one packet-switched communications network, such as an IP based network. The audio signal may also be received from a stereo music server, and/or over a radio network, a fixed tele-
35 communications network, a mobile telecommunications network, a short-range optical link or a short-range radio link.

The step of correcting the erroneous first-channel
data of the first audio channel may involve using the
second-channel data of the second audio channel as well
as preceding non-erroneous first-channel data of the
5   first audio channel.

A second aspect of the invention is a computer
program product directly loadable into a memory of a
processor, where the computer program product comprises
program code for performing the method according to the
10   first aspect when executed by the processor.

A third aspect of the invention is an integrated
circuit, which is adapted to perform the method according
to the first aspect.

A fourth aspect of the invention is a receiver of
15   multi-channel digital audio, comprising

means for receiving an audio signal having audio
data forming a first audio channel and a second audio
channel included therein, said first and second audio
channels being correlated with each other in a manner so
20   that a spatial sensation is typically perceived when
listened to by a user;

means for detecting erroneous first-channel data in
the first audio channel;

means for obtaining second-channel data from the
25   second audio channel; and

means for correcting the erroneous first-channel
data of the first audio channel by using the second-
channel data;

as well as

30   means for determining, upon detection of the er-
roneous first-channel data, a spatially perceivable
inter-channel relation between the first and second audio
channels, wherein

said means for correcting the erroneous first-
35   channel data of the first audio channel is adapted to use
the determined inter-channel relation when correcting the

erroneous first-channel data so as to preserve the spatial sensation perceived by the user.

The receiver may further comprise means for performing the method according to the first aspect.

5   A fifth aspect of the invention is a user terminal for a communications network, the user terminal comprising at least one of an integrated circuit according to the third aspect or a receiver according to the fourth aspect. The communications network may include a mobile

10   telecommunications network, and the user terminal may be a mobile terminal.

The user terminal may be adapted to receive the audio signal from a teleconference bridge over the communications network.

15   A sixth aspect of the invention is a teleconference system comprising a communications network, a plurality of user terminals according to the fifth aspect and a teleconference bridge, wherein the user terminals are connected to the teleconference bridge over the communi-

20   cations network.

Other objectives, features and advantages of the present invention will appear from the following detailed disclosure, from the attached dependent claims as well as from the drawings.

25   Generally, all terms used in the claims are to be interpreted according to their ordinary meaning in the technical field, unless explicitly defined otherwise herein. All references to "a/an/the [element, device, component, means, step, etc]" are to be interpreted

30   openly as referring to at least one instance of said element, device, component, means, step, etc., unless explicitly stated otherwise. The steps of any method disclosed herein do not have to be performed in the exact order disclosed, unless explicitly stated otherwise.

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Brief Description of the Drawings

The present invention will now be described in more detail, reference being made to the enclosed drawings, in which:

FIG 1 is a schematic illustration of a telecommunication system used for transmission of stereo music from a remote server to a mobile terminal, as one example of a case where the present invention may be applied.

FIG 2 is a schematic block diagram illustrating some of the elements of FIG 1.

FIG 3 is a schematic illustration of a teleconference system including a teleconference bridge and a plurality of user terminals, as another example of a case where the present invention may be applied.

FIG 4 is a schematic block diagram of one of the user terminals in FIG 3 according to one embodiment.

FIG 5 is a schematic block diagram of one of the user terminals in FIG 3 according to another embodiment.

FIG 6 illustrates the general error concealment approach according to the invention, where a left stereo channel is used as a source for reconstructing an erroneous right stereo channel together with a determined inter-channel time difference (phase difference between the channels).

FIG 7 is similar to FIG 6 but illustrates the opposite situation, where a right stereo channel is used as a source for reconstructing an erroneous left stereo channel together with a determined inter-channel time difference (phase difference between the channels).

FIG 8 is a flow chart which illustrates the main steps for error concealment according to the invention, when one channel contains an erroneous frame.

FIG 9 is a flow chart which illustrates the main steps for error concealment according to the invention, when both channels simultaneously contain erroneous frames.

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FIG 10 shows a simplified block diagram of an AMR
(Adaptive Multi-Rate) audio decoder.

Detailed Description of the Invention

First, with reference to FIGs 1 and 2, one example
of a multi-channel audio application will be described in
the form of a telecommunication system for transmission
of stereo music from a remote server to a mobile termi-
nal. Then, with reference to FIGs 3-5, another example of
a multi-channel audio application will be described in
the form of a teleconferencing system. The error con-
cealment method according to the invention will be de-
scribed in detail with reference to FIGs 6-10, wherein
the teleconference system of FIGs 3-5 will serve as a
base in a non-limiting manner; the error concealment
method may equally well be applied to the telecommuni-
cation system of FIGs 1-2 as well as in various other
applications not explicitly described herein, as will be
apparent to a skilled person.

* * *

In the telecommunication system of FIG 1, multi-
channel audio in the form of for instance digitally
encoded stereo music may be stored in a database 124 to
be delivered from a server 122 over the Internet 120 and
a mobile telecommunications network 110 to a mobile
telephone 100. To this end, the mobile telephone 100 may
be equipped with a stereo headset 134, through which a
user of the mobile telephone 100 may listen to stereo
music 136 from the server 122. Instead of being stored in
a database 124, the multi-channel audio provided by the
server 122 may be read directly from an optical storage,
such as a CD or DVD. Moreover, the server 122 may be con-
nected to or included in a radio broadcast station so as
to provide streaming audio services across the Internet
120 to the mobile telephone 100. The mobile telephone may
be any commercially available device for any known mobile

telecommunications system, including but not limited to
GSM, UMTS, D-AMPS or CDMA2000.

The system in FIG 1 may be used for audio con-
ferencing. Either the audio conferencing arrangement may
5   be pre-arranged and controlled by the server 122 residing
in the network, as has traditionally been the case, or
the audio conference may be formed as a so-called "ad hoc
conference", wherein one terminal device (e.g. mobile
telephone 100) contacts at least two other terminals and
10   arranges the conference. In the latter case, one of the
terminals may also contain server functionality, and it
may not be necessary to have any network server at all.
Such systems as presented above can be envisioned, e.g.,
in connection with the rich call services provided by the
15   3G and 4G networks.

Of course, multi-channel audio as well as various
other data such as monophonic speech, video, images and
text messages may be communicated between different units
100, 112, 122 and 132 by means of different networks 110,
20   120 and 130. For instance, the portable device 112 may be
a personal digital assistant, a laptop computer with a
GSM or UMTS interface, a smart headset or another acces-
sory for such devices, etc. Moreover, speech may be com-
municated from a user of a stationary telephone 132
25   through a public switched telephone network (PSTN) 130
and the mobile telecommunications network 110, via a base
station 104 thereof across a wireless communication
link 102 to the mobile telephone 100, and vice versa.
FIG 2 presents a general block diagram of a mobile audio
30   data transmission system, including a user terminal 250
and a network station 200. The user terminal 250 may for
instance represent the mobile telephone 100 of FIG 1,
whereas the network station 200 may for instance repre-
sent the base station 104 or the server 122 in FIG 1, or
35   alternatively a teleconference bridge 300 shown in FIG 3.

The user terminal 250 may communicate single-channel
(mono) audio such as speech through a transmission

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channel 206 to the network station 200. The transmission
channel 206 may be provided by the wireless link 102, the
mobile telecommunications network 110 or the Internet 120
in FIG 1, or a packet-switched network 302 in FIG 3, or
5   any such combination. A microphone 252 may receive acous-
tic input from a user of the user terminal 250 and con-
vert the input to a corresponding analog electric signal,
which is supplied to an audio encoding/decoding block
260. This block has an audio encoder 262 and an audio
10   decoder 264, which together form an audio codec. The
analog microphone signal is filtered, sampled and digi-
tized, before the audio encoder 262 performs audio en-
coding applicable to transmission channel 206. An output
of the audio encoding/decoding block 260 is supplied to a
15   channel encoding/decoding block 270, in which a channel
encoder 272 will perform channel encoding upon the en-
coded audio signal in accordance with the applicable
standard for the transmission channel 206.

An output of the channel encoding/decoding block 270
20   is supplied to a radio frequency (RF) block 280, com-
prising an RF transmitter 282, an RF receiver 284 as well
as an antenna (not shown in FIG 2). As is well known in
the technical field, the RF block 280 comprises various
circuits such as power amplifiers, filters, local oscil-
25   lators and mixers, which together will modulate the en-
coded audio signal onto a carrier wave, which is emitted
as electromagnetic waves propagating from the antenna of
the user terminal 250.

After having been communicated across the channel
30   206, the transmitted RF signal, with its encoded audio
data included therein, is received by an RF block 230 in
the network station 200. In similarity with block 280 in
the user terminal 250, the RF block 230 comprises an RF
transmitter 232 as well as an RF receiver 234. The re-
35   ceiver 234 receives and demodulates, in a manner which
is essentially inverse to the procedure performed by the
transmitter 282 as described above, the received RF sig-

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nal and supplies an output to a channel encoding/decoding
block 220. A channel decoder 224 decodes the received
signal and supplies an output to an audio encoding/de-
coding block 210, in which an audio decoder 214 decodes
5   the audio data which was originally encoded by the audio
encoder 262 in the user terminal 250. A decoded audio
output 204, for instance a PCM signal, may be forwarded
within the mobile telecommunications network 110, the
PSTN 130, the Internet 120, the packet-switched network
10   302 in FIG 3, etc. (or to a spatial processing/mixing
unit inside the network station 200, in case it is a
teleconference bridge).

When stereo audio data is communicated in the oppo-
site direction, i.e. from the network station 200 to the
15   user terminal 250, a stereo audio input signal 202 is
received from e.g. the server 122 by an audio encoder 212
of the audio encoding/decoding block 210. After having
applied audio encoding to the audio input signal, channel
encoding is performed by a channel encoder 222 in the
20   channel encoding/decoding block 220. Then, the encoded
audio signal is modulated onto a carrier wave by a
transmitter 232 of the RF block 230 and is communicated
across the channel 206 to the receiver 284 of the RF
block 280 in the user terminal 250. An output of the
25   receiver 284 is supplied to the channel decoder 274 of
the channel encoding/decoding block 270, is decoded
therein and is forwarded to the audio decoder 264 of the
audio encoding/decoding block 260. The audio data is
decoded by the audio decoder 264 and is ultimately con-
30   verted to a pair of analog signals 254, which are
filtered and supplied to left and right speakers for pre-
sentation of the received audio signal acoustically to
the user of the user terminal 250.

As is generally known, the operation of the audio
35   encoding/decoding block 260, the channel encoding/decod-
ing block 270 as well as the RF block 280 of the user
terminal 250 is controlled by a controller 290, which has

associated memory 292. Correspondingly, the operation of
the audio encoding/decoding block 210, the channel en-
coding/decoding block 220 as well as the RF block 230 of
the network station 200 is controlled by a controller 240

5   having associated memory 242.

Even if the audio transmission was described above
as single-channel (mono) from user terminal to network
station but as multi-channel (stereo) in the opposite
direction from network station to user terminal, it is to

10   be understood that this does not necessary always have to
be the case. As an example, mono audio (normal telephone
speech) may for instance be communicated from network
station to user terminal in addition to the stereo audio
referred to above.

15   * * *

In the centralized stereo teleconferencing system
of FIG 3, a plurality of user terminals 310, 320, 330,
340 are connected to the central teleconference bridge
300 through an error-prone network 302, such as a packet-

20   switched IP network. In operation, the teleconference
bridge 300 will receive mono audio streams from the user
terminals 310, 320, 330, 340 and process these mono audio
streams to spatialize them into a stereo output signal
which is supplied to the user terminals. Spatialization

25   can be done e.g. by HRTF filtering the input signals,
thus producing a binaural output signal for each of the
listeners (for headphone listening). The spatialized
stereo audio thus achieved will improve speech intelligi-
bility and facilitate speaker detection and separation,

30   compared to mono teleconferencing. It will also provide a
conference environment sound which is more natural and
satisfactory.

When there is only one active speaker at the same
time, the left and right channels are highly redundant.

35   In theory, if one of these two channels is present, the
other can be reconstructed from the existing one. For
example, a binaural signal produced using HRTF processing

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(linear filtering) satisfies these requirements. However, the exact reconstruction requires knowledge of phase difference and the frequency dependent interaural level difference (ILD) between the channels. The phase diffe-

5   rence is a result of interaural time difference (ITD). ITD varies typically from -0.8 to +0.8 milliseconds corresponding -6 to +6 samples at 8 kHz sampling rate. The ILD is mainly a result of the head shadow effect. The contra-lateral (farther ear) channel has low-pass

10   characteristics compared to the ipsi-lateral (nearer ear).

Reference is now made to FIG 4, which shows one of the user terminals of FIG 3 in more detail. The user terminal 400 has a first interface to the network 302 and

15   is therefore capable of transmitting an encoded mono signal 404 to the teleconference bridge 300. A mono encoder 402 receives audio on a mono channel 420 from the microphone 422 and encodes it into the encoded mono signal 404.

20   The user terminal 400 is also capable of receiving an encoded stereo signal 408 from the teleconference bridge 300. A stereo decoder 406 decodes the signal 408 and forms two decoded channels 438 (left) and 440 (right), which are passed through a mixer 436 and ultima-

25   tely arrive at the left speaker 424 and the right speaker 426. Alternatively, instead of receiving one encoded stereo signal 408 from the teleconference bridge 300, two separately encoded mono channels may be received, one for each stereo channel 438 and 440. As yet an alternative,

30   the left and right channels may have been multiplexed into one common mono channel, in which channels are interleaved.

For successful error concealment with preserved spatial location of the sound source, the user terminal

35   400 needs to identify a phase difference between the stereo channels 438, 400 when reconstructing an erroneous

frame, appearing on one of the channels, from the other
channel.

As previously mentioned, the present invention is
not restricted to the context of audio conferences, but
5   can be used in the reception of any multi-channel audio
signal, e.g., traditional or internet radio transmission,
sound reproduced from media such as CD, minidisc, cas-
sette or MP3 player, or any other memory medium. The re-
ception can take place over a mobile (cellular) network
10   such as GSM, UMTS, CDMA2000 or the like, a local area
network or wide area network such as WLAN or any ad hoc
radio network, or over short-range connectivity like
BlueTooth or other short-range radio connection, or an
optical connection such as IrDA. In such systems, there
15   may not be provided the spatialization information needed
for reconstruction of erroneous or missing data in the
audio signal, and in the following, an embodiment of the
invention is presented where this spatialization informa-
tion is derived by analyzing the received audio signal.
20   In one embodiment the user terminal produces the
required information on phase difference by analyzing the
channels 438, 440. The exact ITD value can be determined
by calculating the cross-correlation between the channels
or by using a phase comparator. Because typically ITD
25   varies between -0.8 and +0.8 ms, the cross-correlation
needs to be calculated only in this window, or if the ITD
sign has already been estimated, calculation can be done
in half of this window. To this end, the energy ratio
between the channels can be used as an estimate for the
30   sign of the ITD value. For example, if there is more
energy on the right channel (interaural level difference
detected), the sound source is more likely to have been
spatialized to the listener's right side which also
defines the sign for the ITD value.
35   An alternative approach, within the above embodi-
ment, is as follows. The left and right channel signals
could first be low-pass filtered with a cutoff frequency

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of, e.g., 400 Hz (the fundamental frequency of speech is
typically below this). The phases of the signals could
then be detected and synchronized during voiced phonemes
by matching peaks and/or zero-crossings. This approach
5   might have an advantage in requiring less computational
power compared to cross-correlation. In yet another
alternative approach, a method like principal-component
analysis (PCA), independent-component analysis (ICA) or
signal-space projection (SSP) may be used to separate the
10   sound sources present in the sound signal. In such
methods, only the strongest or some strongest independent
signal(s) may be separated from the audio signal, and the
correction may be applied for such part(s) only. For
example, in the signal-space projection method, the
15   strongest partial signal is first detected, and its
pattern is removed from the signal. Then the strongest
partial signal left in the audio signal is detected, and
it is subtracted, and so on. This allows for convenient
extraction of a desirable number of prominent audio
20   components from the signal.

Referring back to FIG 4, the stereo decoder 406
will send a frame error indication signal 410 to a
controller 416 whenever a frame on either of the channels
has been lost or corrupted during transmission across the
25   network 302. The controller 416 checks whether the cor-
responding frame on the other channel has been received
correctly. If so, the controller 416 seeks to find the
phase difference between the channels before the error.
This information is provided by a phase & simultaneous
30   speech estimator 412 which is adapted to determine the
phase difference as an ITD value in any of the manners
referred to above. The ITD value is transmitted in a
signal 414 to the controller 416.

The controller 416 controls a multiplexer 432 to
35   select the appropriate one of channels 438 and 440, i.e.
the non-erroneous channel which is to be used for frame
reconstruction of the erroneous channel, to be input to a

spatial reconstruction unit 434. The controller 416 also
derives the ITD value from the signal 414 from the phase
& simultaneous speech estimator 412 and supplies the ITD
value to the spatial reconstruction unit 434.

5   The spatial reconstruction unit 434 will prepare a
frame reconstruction data set in the following manner.
The ITD value is used to determine the first sample of
the audio frame on the unaffected channel, which is re-
ceived through the multiplexer 432 and will be used to
10   replace the erroneous frame of the affected channel. This
is illustrated in more detail in FIG 6, where it is
assumed that the sound source has been spatialized, by
the teleconference bridge 300, at the listener's left
side. A frame error has been determined by the stereo
15   decoder 406 for frame #n in the right channel. Con-
sequently, the phase & simultaneous speech estimator 412
determines that the ITD holds a value of, for instance,
+6 samples, confirming that the phase of the non-
erroneous left channel is ahead of the erroneous right
20   channel. Also fractional samples may be used, but this
requires interpolation during the reconstruction. The
spatial reconstruction unit 434 will receive the deter-
mined ITD value from the controller 416 and prepare the
frame reconstruction data set by copying audio samples
25   starting at 6 samples before the frame boundary of the
concurrent non-erroneous frame #n in the left channel.
The frame reconstruction data set thus prepared has the
same length as the erroneous right-channel frame #n which
it is intended to replace.

30   In FIG 7 a similar situation is shown, where, how-
ever, the erroneous frame #n appears in the leading
channel instead, i.e. in the left channel. In this case,
the frame reconstruction data set is prepared by copying
audio samples starting at 6 samples after the frame
35   boundary of the concurrent non-erroneous frame #n in the
right channel.

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The spatial reconstruction unit 434 will forward the prepared frame reconstruction data set to the mixer 436, optionally after first having performed further processing on the frame reconstruction data set, such as

5   HRTF filtering or adjustment of frequency-dependent ILD level. As already mentioned, the mixer 436 also receives a continuous stream of decoded audio frames on both channels 438 and 440.

The controller 416 controls the mixer 436, through

10   a signal 418, to replace the particular erroneous frame on either of the channels with the corresponding frame reconstruction data set prepared by the spatial reconstruction unit 434, thereby concealing the error to the listener. Thus, corrected stereo audio data arrives at

15   the left and right speakers 424, 426. Optionally, the mixer may do cross-fade between reconstructed and non-erroneous frame boundaries.

The above procedure for concealment of a frame error in one of two stereo channels is illustrated on a

20   more conceptual level in FIG 8. In step 800 it is initially determined that an audio frame error has occurred in one of the channels. In step 802 a phase difference (such as an ITD value) is determined between the non-erroneous channel and the erroneous channel. In step 804

25   it is determined whether the phase difference is positive, i.e. whether the non-erroneous channel is ahead in phase of the erroneous channel. If the answer in step 804 is affirmative, data to be used for frame reconstruction is copied, in an amount corresponding to one audio frame,

30   from the non-erroneous channel in step 806, starting at a certain number of samples before the frame boundary, as illustrated in FIG 6. In the opposite case, in step 808, data is instead copied from the non-erroneous channel starting at a certain number of samples after the frame

35   boundary, as illustrated in FIG 7.

Then, the frame reconstruction data thus prepared may be processed in step 810 in the manners indicated

above. Finally, the erroneous frame on the erroneous channel is replaced by the prepared and processed frame reconstruction data in step 812.

FIG 5 illustrates an alternative embodiment of a
5   user terminal 500 for the teleconference system of FIG 3. Here, the required information on phase difference between audio channels 538 and 540 is derived by the controller 500 directly from metadata 552, which is received together with the encoded stereo signal from the
10 teleconference bridge 300, as indicated at 508. When spatializing the received mono audio channels into stereo audio, the teleconference bridge 300 will include spatial position information of the active sound source (e.g., the current speaker) in the metadata 552. The receiving
15 user terminal 500 will use this spatial position information in the metadata to select the correct ITD value in the error concealment process. In more detail, as one example, the teleconference bridge 300 may use 4 bits to approximate the ITD directly in milliseconds. 1 bit could
20 be used as a sign bit and 3 remaining bits for ITD value in milliseconds, thereby giving an effective ITD range of -0.7 to +0.7 milliseconds in 0.1 ms steps. This information could be assigned to each of the (pairs of) speech frames, or it could be sent more rarely, e.g. with every
25 10th frame only. Whenever frames are lost, the error concealment process uses previously correctly received spatial position information in the error concealment processing.

Thus, there is no need for any local means, such as
30 the phase & simultaneous speech estimator 412 of FIG 4, for estimating the phase difference between the channels from the received audio signal. Other than this, the embodiment of FIG 5 has like components, indicated by like reference numerals, and operates in a manner which
35 is fully equivalent with that of the FIG 4 embodiment.

With reference to FIG 9, an error concealment procedure for a situation with concurrent frame errors in

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both channels is illustrated. As previously mentioned, if audio frames are simultaneously corrupted or missing in both of the channels, to prevent a shift of sound source location, the audio signal for one channel is first

5   reconstructed (from preceding frames in that channel), and then the other channel is reconstructed based on the first reconstructed channel in the manner described above.

Starting with step 900 it is determined that audio

10   frame errors have occurred in simultaneous frames for both channels. In step 902 a determination is made as to which channel to use as source for initial frame re-construction. This determination may be made by investi-gating the signal energy or power level of the two

15   channels and then selecting, as source channel, the one of the channels which has the highest signal energy or power level. Alternatively, the phases of the two channels may be determined, wherein the one of the channels which has leading phase is selected as source

20   channel.

In step 904 the erroneous frame of the selected source channel is reconstructed from preceding correctly received frames of that channel. There are known methods of such intra-channel frame reconstruction which may be

25   used in step 904. Extrapolation by attenuated copying of previous frames is one example.

Then, in step 906, the concurrent erroneous frame of the other channel is reconstructed from the just re-constructed source channel frame in the manner described

30   above and illustrated in FIG 8.

Optionally, as indicated with dashed lines in FIG 9, the quality of the reconstructed output signals can be further improved by controlling the signal gain in the mixer 436/536. After the reconstruction, both channels

35   could be attenuated gradually down to e.g. -10 dB during the first erroneous frame, as shown in step 908. The level of the signals is then kept low for consecutive

frame errors, until it is determined, in step 910, that there are no more consecutive erroneous frames to be corrected. Upon this determination, the first non-erroneous frame in each channel is amplified gradually

5   back to a 0 dB level, as seen in step 912. This option is particularly useful when frames have been lost in both channels, but it may also be applied to the error con-cealment of single-channel errors illustrated in FIG 8.

<p style="text-align:center">* * *</p>

10   An alternative embodiment of error concealment with preserved spatial sensation according to the invention will now be described. This alternative embodiment suggests a modification or extension of the typical intra-channel error concealment methods of contemporary

15   speech codecs so as to make use also of audio data on the other channel for reconstructing erroneous audio frames. In this alternative embodiment, error concealment occurs in "parameter" domain rather than time domain.

     First, some high-level principles of speech com-

20   pression and error concealment will be introduced to better illustrate the fundamentals of this embodiment. In modern speech codecs, such as 3GPP AMR (Adaptive Multi-Rate), the encoder transforms the input speech into a set of parameters that describe the contents of the current

25   frame. These frames are transmitted to the decoder, which uses the parameters to reconstruct a speech signal sounding as closely as possible like the original signal. For example, the parameters transmitted by the AMR codec for each frame are a set of line spectral frequencies

30   (LSFs) for forming the LP synthesis filter, pitch period and pitch gain for the adaptive codebook excitation, and pulse positions and gain for the fixed codebook exci-tation.

     FIG 10 shows a simplified block diagram of an AMR

35   decoder 1000. The adaptive codebook excitation 1010 is formed by copying the signal from the adaptive codebook 1002 from the location indicated by the received pitch

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period, and multiplying this signal with the received
pitch gain, as seen at 1006. The fixed codebook exci-
tation 1012 for the fixed codebook 1004 is built based on
received pulse positions and by multiplying this signal
5   with the received fixed codebook gain, as seen at 1008.
The sum 1014 of adaptive codebook and fixed codebook
excitations 1010, 1012 forms the total excitation 1016,
which is processed by an LP synthesis filter 1018, formed
based on the received LSFs, to reproduce a synthesized
10   speech signal 1020. Furthermore, the total excitation
1016 is also fed back, at 1022, to the adaptive codebook
memory to update the adaptive codebook 1002 for the next
frame.

       Since, in short term, the speech signal is quite
15   stationary in nature (in terms of energy level and
spectral content), also many of the parameters used to
describe the signal will evolve relatively slowly over
time. While this short-term stationarity is one of the
fundamentals of efficient compression that exploits
20   intra-channel, inter-frame dependency, it also enables
quite efficient error concealment techniques simply by
extrapolating the parameter values based on their values
in previous frame(s) in the same channel. An example
solution for the error concealment for the AMR codec is
25   thoroughly described in "3GPP TS 26.091 AMR speech codec;
Error concealment of lost frames (Release 4), version
4.0.0 (2001-03)".

       Since the error concealment is performed in
"parameter domain" (instead of modifying the signal in
30   time domain), this will also be a computationally effi-
cient operation, which can be performed generally by
using the same algorithms as normal speech decoding. The
general principle of smooth error concealment is to avoid
annoying sounds by gradually downscaling the signal
35   energy and forcing the spectrum more and more flat by
modifying the parameter values by predefined factors.
However, despite the assumed stationarity, the parameters

are naturally gradually changing over time, and with in-
creasing number of consecutive missing frames the result
of error concealment gets worse and worse.

For instance, in case of a lost frame, the example
5   approach to error concealment in the AMR codec computes
the LSF parameters by shifting the LSF values from the
previous frame slightly towards their means, resulting in
".flatter" frequency envelope. The pitch period is either
directly copied or slightly modified from the previous
10 frame. For pitch gain and fixed codebook gain slightly
adjusted ("downscaled") values are used, based on the few
most recently received values.. The pulse positions of
the fixed codebook excitation are not assumed to have
dependency between successive frames (on the same
15 channel), and the error concealment procedure can select
them randomly. However, with increasing number of con-
secutive missing frames the "downscaling" factor for
pitch gain and fixed codebook gain is increased, re-
sulting eventually in total muting of the decoder output
20 after five or six missing frames.

In view of the above, the aforesaid alternative
embodiment of the invention proposes two different error
concealment scenarios for two-channel spatial speech.

A.   Frame missing only from one channel: Since in a
25 spatialized stereo teleconference application or a stereo
music application the two channels are highly correlated,
the parameters received in the frame for the other
channel can be used to enhance the error concealment
performance on the channel where the frame is missing.
30 Even if there is a small phase difference between the
channels (in the range from -6 to +6 samples, as de-
scribed earlier), when the parameters of a frame are
evaluated e.g. over 160 samples (corresponds to 20 ms
frame length at 8 kHz sample rate), parameter estimation
35 based on the other (non-erroneous) channel will give a
better approximation of the real parameter values than
the ones that have been extrapolated within the channel

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with the erroneous frame. Thus, error concealment will
work better when parameter information from the other
channel can be used in addition to normal extrapolation-
based parameter estimation. For example, in standard
5   error concealment for the AMR codec the pitch gain and
codebook gain are downscaled based on previously received
values according to a predefined pattern (reference is
made to the AMR specification referred to above for
details). This has proven to be a good and safe solution
10   for a single-channel case, but it will not give optimum
performance for a spatialized two-channel case.

For example, consider a situation where these gain
values would be changing towards larger values in the
frame that has been lost or corrupted: the standard AMR
15   error concealment would downscale the signal in the
erroneous channel, while in the other channel the signal
level would go up according to the actual data in the
correctly received frame, thus generating a clear dif-
ference between the channels. As a result, the spatial
20   image would move to a "wrong" position. Thus, the in-
vention proposes using parameter information received for
the other channel to enhance the error concealment per-
formance of the erroneous channel by indicating the
correct "trend" of the change of the signal characte-
25   ristics (e.g. scaling signal value up instead of down).
This would yield better speech quality.

Improved two-channel error concealment performance
could be reached by directly copying the LSFs from the
non-erroneous channel, or adjusting these values slightly
30   towards values computed by the "normal" error concealment
procedure for the erroneous channel. Similarly, the pitch
period, pitch gain, and fixed codebook gains for the
erroneous channel can be taken from the non-erroneous
channel, either directly or modified with a scaling
35   factor. The scaling factor could be adaptive in such a
way that its value is constantly updated to be the ratio
between the parameter (i.e. pitch period, pitch gain, or

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fixed codebook gain) values in both channels. Further-
more, the scaling factor could also take into account the
parameter value history of the erroneous channel. Al-
though it is considered to be sufficient to just ran-

5   domize the fixed codebook excitation pulse positions,
this might cause a phase difference for a two-channel
spatialized signal, especially during unvoiced speech
where the fixed codebook typically provides the major
contribution to the total excitation. Therefore, in a

10   two-channel case, the error concealment performance would
be improved if the pulse positions are copied from the
non-erroneous channel and shifted according to the ITD
before forming the total excitation used in the erroneous
channel.

15   B.   Frame missing from both channels: When frames
from both channels are lost, there is naturally no re-
dundant information to be used to enhance the actual
error concealment. However, also in this case knowledge
of phase and energy difference between channels can be

20   used to improve speech quality by preserving the spatial
position also in the extrapolated signal. Either of the
channels is simply selected as the "source channel" (for
instance the one with higher energy level), and the error
concealment is performed for this channel as in the stan-

25   dard single-channel case. After this, the extrapolated
frame is regarded as if it were a normally received
frame, and the error concealment is performed as de-
scribed above in case A. This approach makes sure that
the concealment on both channels changes the parameter

30   values according to a similar pattern, thus minimizing
the deviation between the channels that might shift the
spectral position.

                          * * *

     Various alternatives may be applied to the em-
35   bodiments described above. Some of those alternatives
will be briefly mentioned below, in a non-exhaustive
manner.

30

As regards the phase difference between the
channels, one possibility to improve the detection of ITD
is to analyze the spectrum of signals in bands and take
advantage of frequency dependency of ILD. In HRTF pro-
5   cessed signals the ITD value correlates with the effect
of head shadow, which has low-pass filter characteris-
tics. Thus, a sound source that is at the right side of
the listener (positive ITD) has less high frequency
energy in the left channel (farther ear) than in the
10   right channel (nearer ear). The more the high frequencies
of the signal are attenuated at the farther ear side
comparing to the nearer ear side, the farther the sound
source is spatialized from the center. This method re-
quires knowledge of the spatialization algorithm used in
15   the teleconference bridge.

When reconstructing an erroneous audio frame in the
spatial reconstruction unit 434, it is possible to use
different filters depending on the spatial position of
the sound source and the reconstruction direction, i.e.
20   whether it is from contra-lateral to ipsi-lateral or vice
versa. If the contra-lateral channel is lost, it can be
generated by low-pass filtering of the ipsi-lateral
channel. Correspondingly, the ipsi-lateral channel can be
generated by boosting the high frequencies of the contra-
25   lateral channel. This approach requires knowledge of the
spatialization algorithm used in the teleconference
bridge 300.

In case of simultaneous audio frame errors on both
channels, if frames are lost or corrupted in the middle
30   of a voiced sound, it might be useful to replace a few
consecutive correct frames after the last erroneous
frame. When the decoder extrapolates a lost frame (e.g.
step 904 in FIG 9), it automatically attenuates output
signal level. Correspondingly, when the next non-
35   erroneous frame is decoded, output signal level is
gradually amplified to the target level. This can cause
discontinuity in the amplitude envelope at the border

between a reconstructed frame and the following non-erroneous frame, which can be heard as a click. To overcome this problem extra frames could be processed.

Additionally, if there are more than 6 missing
5   frames at both channels, it might not be necessary to process the additional frames exceeding this number. The decoder would already have attenuated the signal level during extrapolation down to a mute level.

The error concealment method of the invention works
10   also for binaural recordings or speech that is captured from a conference room by two microphones.

Moreover, it can be applied to a stereo codec or dual mono codecs as well, such as the audio encoding/decoding block 260 indicated in FIG 2 which may
15   be a MPEG-4 or MPEG-2 AAC (Advanced Audio Coding) codec, an ISO/MPEG Audio Layer-3 (MP3) codec, or two mono codecs such as GSM EFR/FR/HR speech codec, AMR, Wideband AMR, G.711, G.722, G.722.1, G.723, G.728, or according to MPEG1/2/4 CELP+AAC codec. If a frame has been lost during
20   transmission, the extrapolated signal can be spatialized to the correct location at the terminal using the method according to the invention. The error concealment method could be applied in a stereo codec for transmitting spatial speech. The error concealment method would extra-
25   polate, in addition to signal waveform, the spatial position. The presented method could be integrated in a stereo codec which allows to specify the content of the signal as a meta information. The method would be taken into use whenever it is specified that the signal is
30   spatialized speech.

The presented error concealment method works best if room effect (reverb) is added to the spatialized signals in the terminal after the error concealment processing. If the room effect is processed already in
35   the teleconference bridge, the error concealment at the terminal spatializes also the reverb energy, which is supposed to be diffuse and non-spatial from the lis-

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tener's aspect, to the same spatial position in which the
sound source is localized. This may degrade the spatial
audio quality a bit, because the feeling of audio im-
mersion degrades . However, because the error concealment
5   works at short time scale (typically 20-200ms), this
might not be a noticeable problem in most cases. In
addition, when the room effect is added in the terminal
it can even mask some anomalies that are generated in the
error concealment process.

10   The error concealment functionality described above
may be realized as an integrated circuit (ASIC) or as any
other form of digital electronics. In an alternative
embodiment, the error concealment functionality may be
implemented as a computer program product, which is
15   directly loadable into a memory of a processor. The
processor may be any CPU, DSP or other kind of micro-
processor which is commercially available for personal
computers, server computers, palmtop computers, laptop
computers, etc, and the memory may be e.g. RAM, SRAM,
20   flash, EEPROM and/or an internal memory in the processor.
The computer program product comprises program code for
providing the error concealment functionality when exe-
cuted by the processor.

           It is to be emphasized, again, that the invention
25   is not limited to two channels but may be applied to an
arbitrary number of channels in excess of a single
channel. For instance, the invention could be applied to
a 4.1, 5.1 or 6.1 digital audio format, or any other so-
called 3D or spatial audio format, or in general any two
30   channels which carry audio information and are temporally
highly correlated, i.e. derived essentially from the same
sound source.

           The invention could be extended into a case where
ITD detection is done separately for each sub-band
35   between the input signals. As a result, an estimate of
spatial position of the sound source at each sub-band
will be detected. When frame loss happens, in the error

concealment processing all these positions would be preserved separately. This method would suit multi-speech signals and music. To this end, a method of detecting the location of a sound source is described in Liu, C.,

5   Wheeler, B. C., O'Brien, W. D., Bilger, R. C., Lansing, C. R., and Feng, A. S. "Localization of multiple sound sources with two microphones", J. Acoust. Soc. Am. 108 (4), pp. 1888-1905, Oct. 2000, which is incorporated herewith by reference.

10   The invention has mainly been described above with reference to a few embodiments. However, as is readily appreciated by a person skilled in the art, other embodiments than the ones disclosed above are equally possible within the scope of the invention, as defined by

15   the appended patent claims.

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CLAIMS

1. An error concealment method for multi-channel digital audio, the method comprising the steps of

receiving an audio signal having audio data forming

5   a first audio channel and a second audio channel included therein, said first and second audio channels being correlated with each other in a manner so that a spatial sensation is typically perceived when listened to by a user;

10   detecting erroneous first-channel data in the first audio channel;

obtaining second-channel data from the second audio channel; and

correcting the erroneous first-channel data of the

15   first audio channel by using the second-channel data;
characterized by

determining, upon detection of the erroneous first-channel data, a spatially perceivable inter-channel relation between the first and second audio channels; and

20   using the determined inter-channel relation when correcting the erroneous first-channel data of the first audio channel so as to preserve the spatial sensation perceived by the user.

25   2. A method as in claim 1, wherein the erroneous first-channel data of the first audio channel is corrected by manipulating the second-channel data in accordance with the determined inter-channel relation and then replacing the erroneous first-channel data with the

30   manipulated second-channel data.

3. A method as in claim 1 or 2, wherein the determined inter-channel relation is a phase difference between the first and second audio channels.

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4. A method as in claim 2 and 3, wherein the manipulation of the second-channel data consists in selecting

said second-channel data from the second audio channel
with a time shift with respect to the first audio
channel, said time shift corresponding to the determined
phase difference.

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5. A method as in claim 3, wherein the phase diffe-
rence is determined by analyzing the first and second
channels of the received audio signal with respect to
each other.

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6. A method as in claim 5, wherein the analysis of
the first and second channels of the received audio sig-
nal involves calculating the cross-correlation between
the channels.

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7. A method as in claim 5, wherein the analysis of
the first and second channels of the received audio sig-
nal involves:
low-pass filtering of each of the first and second
20   channels; and
detecting the phases of the first and second
channels after low-pass filtering by matching peaks
and/or zero-crossings in voiced phonemes.

25   8. A method as in claim 1, wherein the spatially
perceivable inter-channel relation is determined from
metadata received together with the audio signal.

9. A method as in any preceding claim, comprising
30   the additional step of decoding the received audio signal
prior to said step of detecting erroneous first-channel
data in the first audio channel.

10. A method as in any preceding claim, wherein the
35   first and second audio channels each comprises a
plurality of audio frames and wherein the detection and

correction of erroneous first-channel data concern at
least one entire audio frame.

11. A method as in any of claims 1-9, wherein the
5   first and second audio channels each comprises a
plurality of audio frames and wherein the detection and
correction of erroneous first-channel data concern
part(s) of an audio frame.

10   12. A method as in claim 11, wherein said part(s)
of an audio frame relate(s) to spectral sub-band(s).

13. A method as in claim 10 or 11, wherein the
detection and correction of erroneous first-channel data
15   is performed on a plurality of time domain audio samples
contained in the audio frame.

14. A method as in any preceding claim, wherein the
first and second audio channels are left and right stereo
20   channels, or vice versa.

15. A method as in claim 1, comprising the addi-
tional steps, after said step of detecting erroneous
first-channel data in the first audio channel, of:
25   detecting erroneous second-channel data in the
second audio channel, essentially concurrent with the
erroneous first-channel data detected in the first audio
channel;
      selecting either the first audio channel or the
30   second audio channel as source channel for audio re-
construction;
      reconstructing the erroneous data of the selected
source channel from preceding data in the selected source
channel; and
35   reconstructing the erroneous data of the other of
the first and second audio channels, which was not
selected as source channel, from the reconstructed data

of the source channel in accordance with the remaining
steps of claim 1.

5   16. A method as in claim 15, wherein the one of the
first audio channel or the second audio channel which has
the highest signal energy or power level is selected as
source channel.

        17. A method as in claim 15, wherein the one of the
10   first audio channel or the second audio channel which is
leading in terms of phase is selected as source channel.

        18. A method as in claim 17, wherein said step of
reconstructing the erroneous data of the selected source
15   channel from preceding data in the selected source
channel is performed by attenuated extrapolation or
copying of said preceding data.

        19. A method as in any preceding claim, wherein the
20   audio signal is received over at least one packet-
switched communications network.

        20. A method as in any preceding claim, wherein the
audio signal is received from a teleconference bridge.
25

        21. A method as in any preceding claim, wherein the
audio signal is received from a stereo music server.

        22. A method as in any preceding claim, wherein the
30   audio signal is received over a radio network, a fixed
telecommunications network, a mobile telecommunications
network, a short-range optical link or a short-range
radio link.

35   23. A method as in any preceding claim, wherein the
first and second audio channels are any two correlated
channels of a 4.1, 5.1 or 6.1 digital audio format.

38

24. A method as in any one of claims 15-18, comprising the additional steps of

attenuating the reconstructed data of said source

5   channel and said other channel;

maintaining the first and second audio channels attenuated for as long as there are consecutive errors on the first and second audio channels; and

upon detecting that there are no more consecutive

10   errors on the first and second audio channels, amplifying the first and second audio channels to cancel the attenuation thereof.

25. A method as in claim 1, wherein the step of

15   correcting the erroneous first-channel data of the first audio channel involves using the second-channel data of the second audio channel as well as preceding non-erroneous first-channel data of the first audio channel.

20   26. A method as in claim 25 and having the step of decoding according to claim 9, wherein the received audio signal is decoded by at least one codec, such as an MPEG-4 or MPEG-2 AAC codec, an ISO/MPEG Audio Layer-3 (MP3) codec, or two mono codecs like GSM EFR/FR/HR speech

25   codec, AMR, Wideband AMR, G.711, G.722, G.722.1, G.723, G.728, or an MPEG1/2/4 CELP+AAC codec.

27. A method as in claim 1, wherein the detection and correction of erroneous first-channel data concern

30   audio component(s), such as principal audio component(s), which is/are detected or indicated to be present in the audio signal.

28. A computer program product directly loadable

35   into a memory of a processor, where the computer program product comprises program code for performing the method

according to any of claims 1-27 when executed by said
processor.

29. An integrated circuit, which is adapted to
perform the method according to any of claims 1-27.

30. A receiver of multi-channel digital audio,
comprising
    means for receiving an audio signal having audio
data forming a first audio channel and a second audio
channel included therein, said first and second audio
channels being correlated with each other in a manner so
that a spatial sensation is typically perceived when
listened to by a user;
    means for detecting erroneous first-channel data in
the first audio channel;
    means for obtaining second-channel data from the
second audio channel; and
    means for correcting the erroneous first-channel
data of the first audio channel by using the second-
channel data;
    characterized by
    means for determining, upon detection of the
erroneous first-channel data, a spatially perceivable
inter-channel relation between the first and second audio
channels, wherein
    said means for correcting the erroneous first-
channel data of the first audio channel is adapted to use
the determined inter-channel relation when correcting the
erroneous first-channel data so as to preserve the
spatial sensation perceived by the user.

31. A receiver as in claim 30, further comprising
means for performing the method according to any of
claims 1-27.

40

32. A user terminal (100; 310) for a communications network (110, 120; 302), the user terminal comprising at least one of an integrated circuit according to claim 29 and a receiver according to claim 30 or 31.

5

33. A user terminal as in claim 32, wherein the communications network (110) includes a mobile telecommunications network and the user terminal is a mobile terminal (100).

10

34. A user terminal as in claim 32, adapted to receive the audio signal from a teleconference bridge (300) over the communications network (302).

15   35. A teleconference system (300) comprising a communications network (302), a plurality of user terminals (310, 320, 330, 340) according to claim 32 and a teleconference bridge (300), wherein the user terminals are connected to the teleconference bridge over the
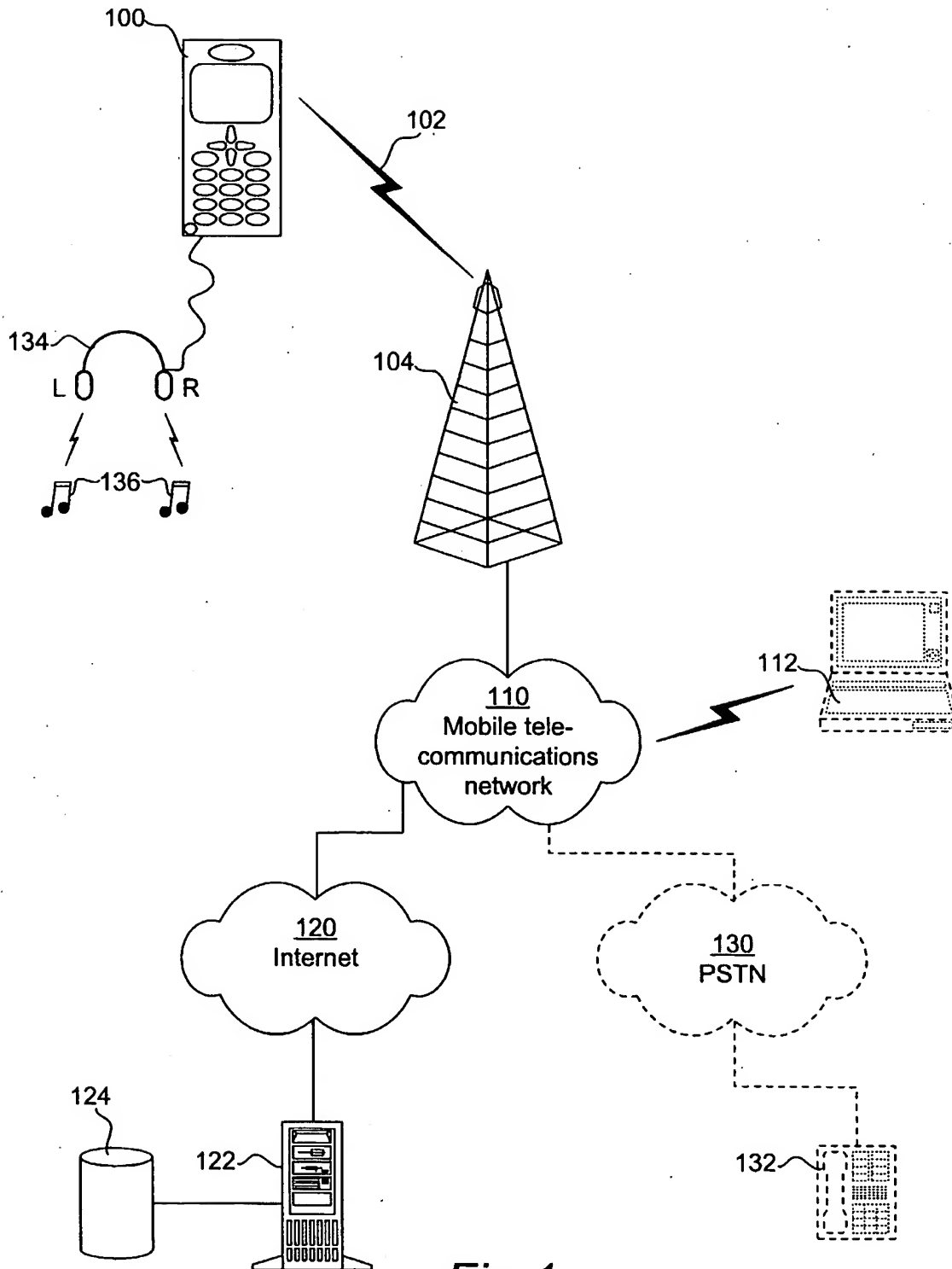
20   communications network.
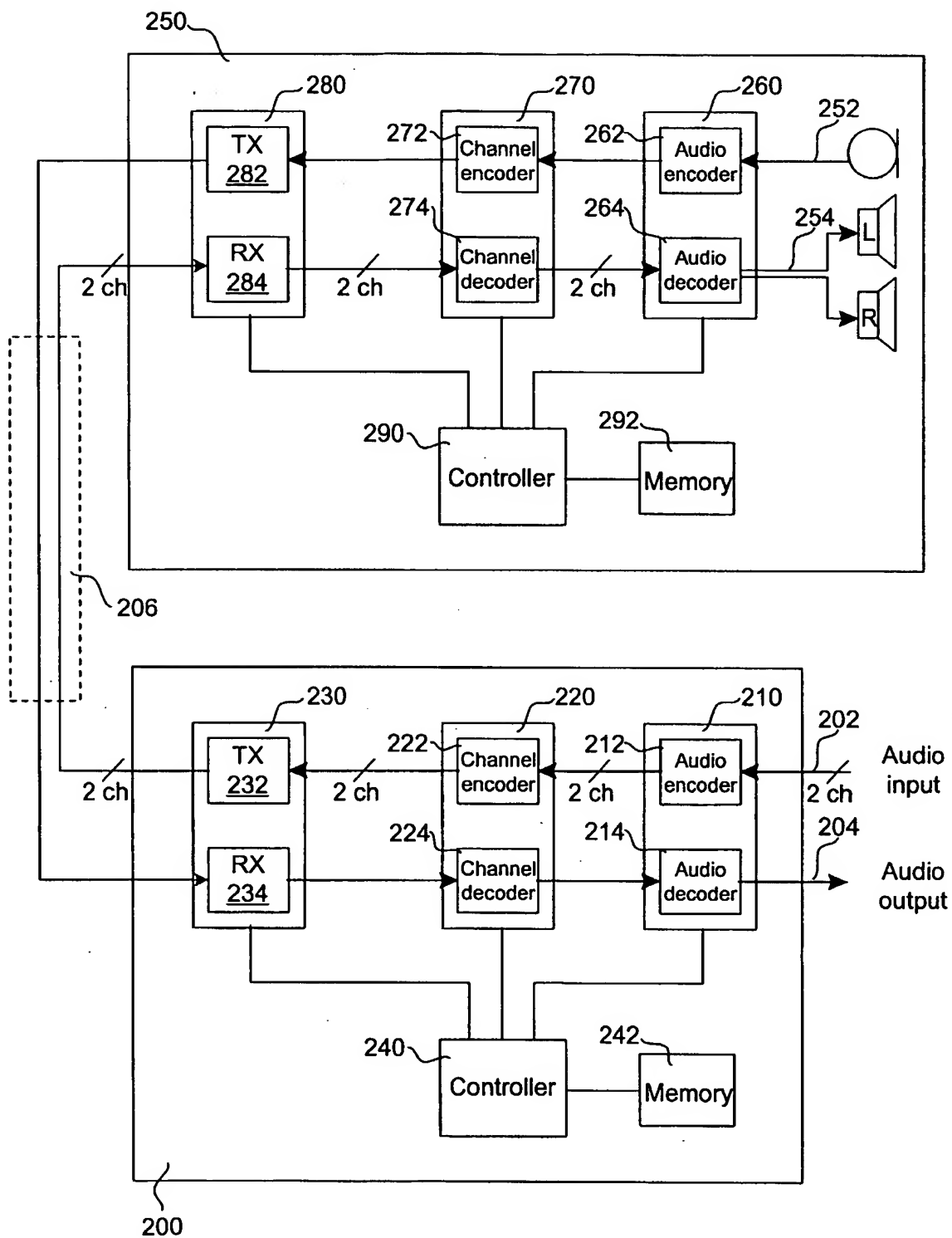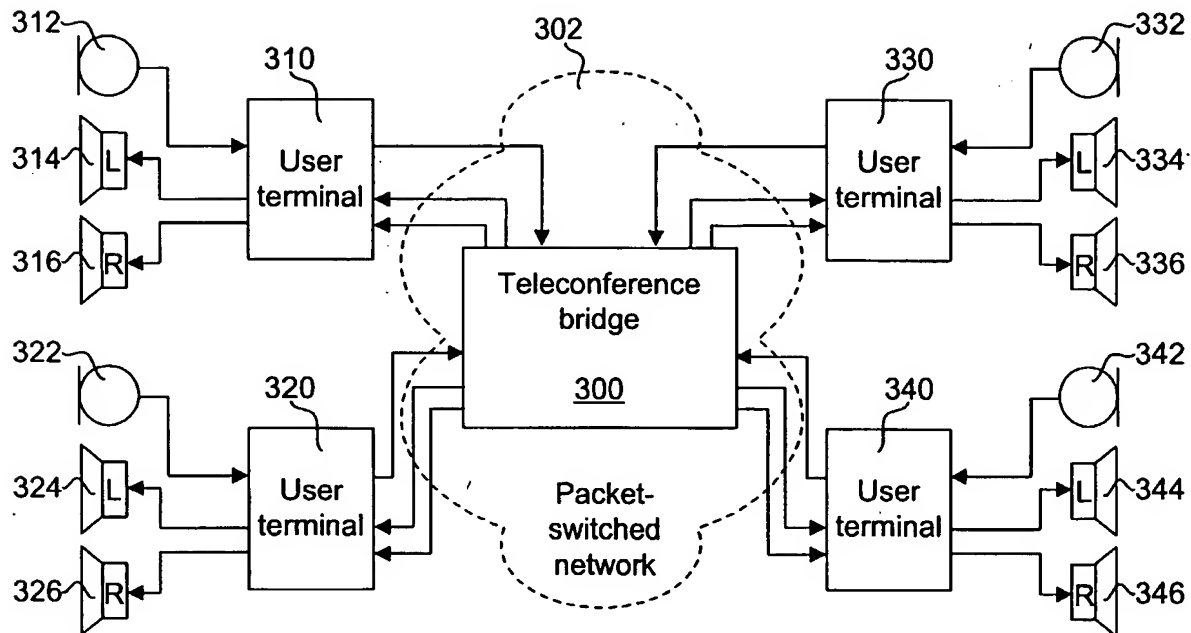
Fig 1

Fig 2

Fig 3

Fig 4

Fig 5

Fig 6

Fig 7

5/7

Error concealment
Frame error in one channel

800 — Detect frame error

802 — Determine phase difference (ITD) between non-erroneous and erroneous channels

804 — Phase difference positive?

Yes   No

806 — Copy frame from non-erroneous channel, starting at ITD No of samples before frame boundary

808 — Copy frame from non-erroneous channel, starting at ITD No of samples after frame boundary

810 — Process copied frame (filtering, ILD adjustment)

812 — Replace erroneous frame with processed frame

Fig 8

Error concealment
Frame error in both channels

900 — Detect frame errors in both channels

902 — Determine which channel to use as source for reconstruction

904 — Reconstruct current frame in selected source channel by extrapolating preceding frames in this channel

906 — Reconstruct current frame in other channel from reconstructed current frame in source channel channel

908 — Attenuate first frame in both channels gradually

910 — More consecutive frame errors?

Yes

No

912 — Amplify first non-erroneous frame in both channels gradually

Fig 9

Fig 10

A. CLASSIFICATION OF SUBJECT MATTER

IPC7: H04L 1/00, H04H 1/00

According to International Patent Classification (IPC) or to both national classification and IPC

B. FIELDS SEARCHED

Minimum documentation searched (classification system followed by classification symbols)

IPC7: H04L, H04B, G11B, G10L, H04H

Documentation searched other than minimum documentation to the extent that such documents are included in the fields searched

SE,DK,FI,NO classes as above

Electronic data base consulted during the international search (name of data base and, where practicable, search terms used)

EPO-INTERNAL, WPI DATA, PAJ, INSPEC

C. DOCUMENTS CONSIDERED TO BE RELEVANT

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A	EP 0738056 A2 (GRUNDIG E.M.V.), 16 October 1996 (16.10.96), figures 1,3, abstract --	1-35
A	PATENT ABSTRACTS OF JAPAN Vol. 012, No. 457 (E-688) 30 November 1988 (1988-11-30) abstract & JP 63 182977 A (PIONEER ELECTRONIC CORP) 28 July 1988 (1988-07-28) --	1-35
A	US 6351727 B1 (WIESE, D. ET AL), 26 February 2002 (26.02.02), column 1, line 42 - line 59; column 3, line 42 - line 52 --	1-35

| X | Further documents are listed in the continuation of Box C. | | X | See patent family annex. |

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"&"	document member of the same patent family

Date of the actual completion of the international search	Date of mailing of the international search report
19 November 2003	2 5 -11- 2003

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C (Continuation). DOCUMENTS CONSIDERED TO BE RELEVANT		
Category*	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
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Information on patent family members

30/12/02

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